

# ESD Design in High Voltage Smart Power Technologies

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# Acknowledgments

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# Instructor's biography

**Lorenzo Cerati** is ESD Senior Principal Engineer and senior member of the Technical Staff at STMicroelectronics. Lorenzo received his M.Sc. degree in telecommunication engineering at the "Politecnico di Milano" Technical University in 1998, discussing a thesis on a CdZnTe cross-connect for optical networks. Since 2000, he has been working in STMicroelectronics in the ESD protections development team for Smart power technologies and now he is the manager of the group responsible for ESD protections development, Latch-up immunity and bipolar parasitic analysis in BCD processes. He is also in charge of design teams support to define, implement and debug complex ESD architectures in BCD ICs.

Lorenzo represents STMicroelectronics in the ESDA standardization committees and is member of JWG HBM, JWG CDM, WG5.4, WG 5.5 (where he is acting as vice-chairman), WG 5.6, and WG 14. Starting from 2016 he also has been serving as member of the Technical and Advisory Support Committee (TAS). Lorenzo served multiple times as a member of the ESREF, ICICDT, IRPS, IEW, and EOS/ESD Symposium Technical Program Committees and authored several papers on ESD and EDA topics. Since 2015, Lorenzo is member of the EOS/ESD symposium steering committee and will be Vice General Chair at the 2019 EOS/ESD symposium. Lorenzo is member of the ESDA Board of Directors' since 2015.

# Abstract

ESD design in High Voltage Smart Power is a very challenging task due to the peculiarities of these technology nodes. After a short introduction describing the status of the international standards and norms for the various application fields where these technologies are adopted, an overview of the process options and the typical device portfolios in the Smart Power world will be given. Different ESD protection concepts will be introduced, analyzing advantages and disadvantages of the various possible approaches to implement ESD networks (diodes, snapback, active clamps...). Finally, some examples of HV-technology and design-related challenges regarding ESD protections will be discussed, with a focus on parasitic bipolar formation and their impact on device performance.

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- Part 1 - Introduction: International ESD standards and norms
- Part 2 - Introduction to Smart Power Technologies & device portfolio
- Part 3 - ESD Protection concepts in Smart Power technologies
- Part 4 - HV related ESD issues

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- Part 1 - Introduction: International ESD standards and norms
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# What is an ESD?

- An Electrostatic Discharge is a rapid discharge event that transfers a finite amount of charge between two bodies at different potentials
- The shape of the current is governed by the dynamic impedance of the two bodies
- The ESD discharge derives from an external event and is statistical in nature
- ESD events are one of the possible root causes for IC's failure on the field

# How to face the problem

- In order to minimize ESD impact we can follow a twofold approach:
  - adopt static-safe practices to control charge build-up
  - produce robust ICs through process and circuit improvements

# The ESD problem in IC

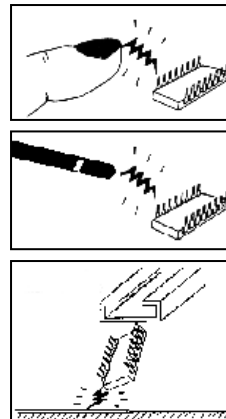
- ESD events can seriously damage ICs

- The charge transfer is due to:

- ☐ Human handling (HBM)

- ☐ Machine handling (MM)

- ☐ Self-charging (CDM)

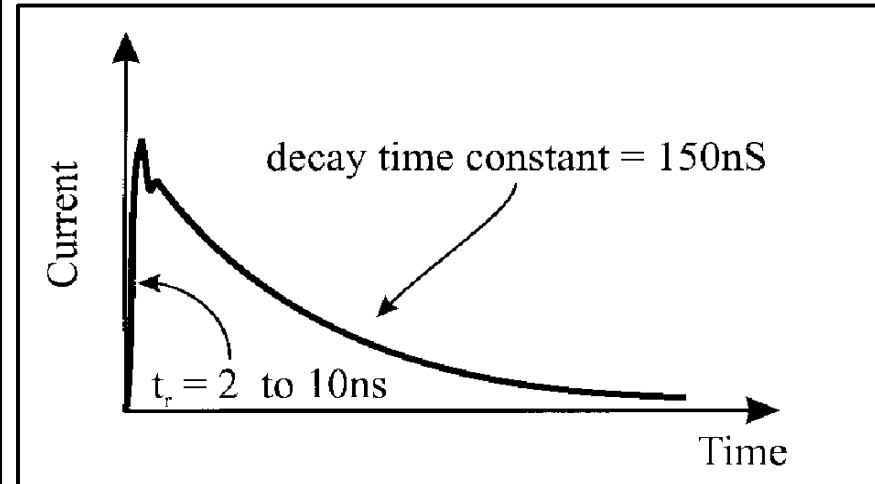
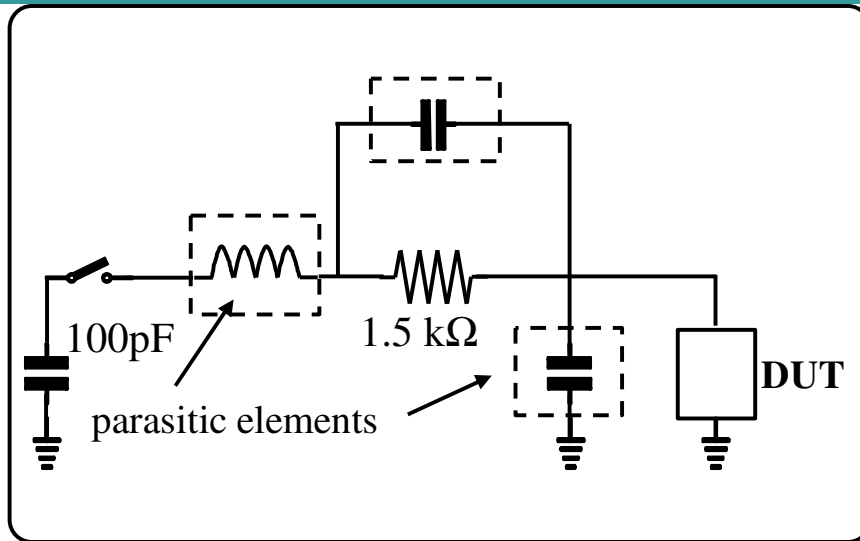


- High current and high voltage are typical for and ESD event
- On-chip protections are mandatory in order to preserve chip functionality

# Charge transfer modeling

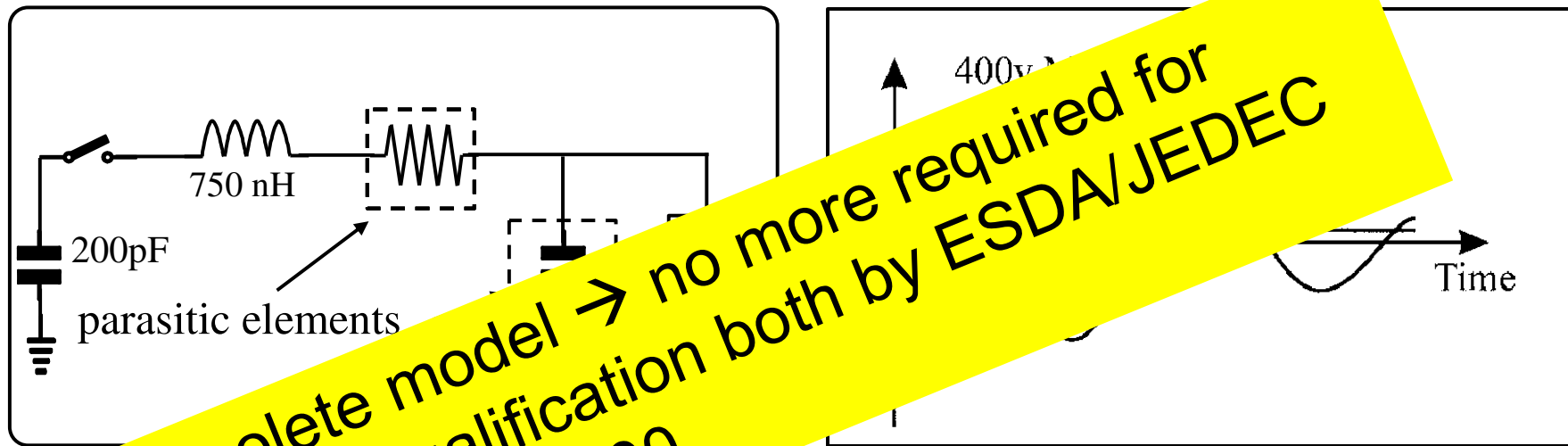
- ESD voltage levels vary drastically with atmospheric conditions
- A source of known and controlled ESD is needed in order to reproduce the discharge in lab → ESD simulator
- Stress level must be the same for any and all commercial ESD simulators
- ESD simulators should be able to reproduce the characteristics of the charge transfer processes which occurs in the real world
- Three different models have been developed
  - Human body model (HBM)
  - Machine Model (MM)
  - Charged device model (CDM)

# HBM: Human Body Model



- The Human Body Model attempts to reproduce an ESD waveform generated by the discharge of a human being through a low impedance path
- Rise time  $< 10 \text{ ns}$ , decay time @  $150 \text{ ns}$

# MM: Machine model

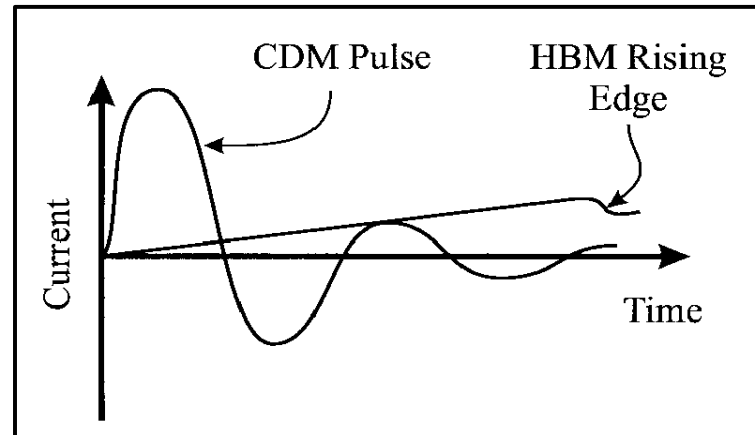


Obsolete model → no more required for product qualification both by ESDA/JEDEC and by AEC-Q100

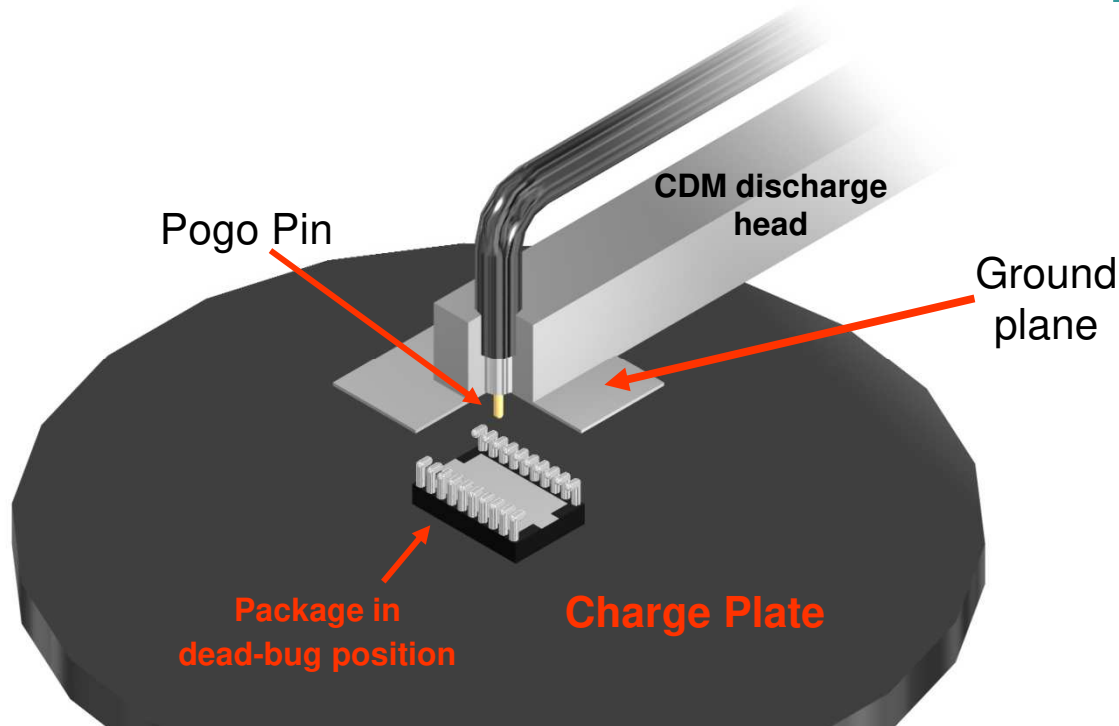
- The Machine Model attempts to reproduce an ESD waveform generated by the discharge of a charged machine
- It reaches higher current levels for a given discharge voltage compared to HBM

# CDM: Main characteristics

- Discharge time:  $\approx 0.5/1\text{ns}$  ( $t_r \approx 100/400\text{ps}$ )
- High current peak: till 20A for 1kV pre-charge
- Complete IC involved:
  - Full IC protection strategy must be implemented
  - Internal nodes could be critical (coupling capacitances)
- Stress dependent on parasitic elements
  - Package
  - Bond wires
  - Bond pads
  - Substrate
  - Well diodes
- High current injection phenomena → Forward recovery effect



# CDM: tester description



- Package in dead bug position
- Pre-charged by charge plate induction  
→ Mobile charges are on the device only
- To create the ESD event a probe touches down one pin of the device.
- Minimum RLC circuit simulating real word conditions.
- L & C of the device impact on the current waveform

# Device vs. System level tests: two Worlds of ESD



## Device level

- IC's can be exposed to ESD during assembly
- ESD on-chip protection is used to protect IC during manufacturing
- Qualification of IC by (JEDEC / ESDA / AEC) standards:
  - Human Body Model (HBM)
  - Machine Model (MM)
  - Charged Device Model (CDM)
- Stresses are applied to all pins of the IC



## System level

- ESD is also a threat for PCBs and (final) products
- Dedicated diodes are added on the PCB in order to survive such threats
- System level ESD standards:
  - IEC 61000-4-2
  - ISO 10605
- Stresses are applied to specified locations of the application,  
BUT: Stressing to component pins is prohibited!

# Contents

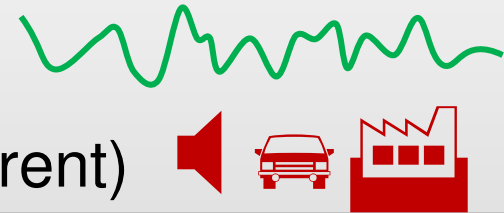
- Part 1 - Introduction: International ESD standards and norms
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# Introduction to HV technologies

## HV technologies combine:

- Digital signal processing
- Analog signal processing
- Power (high voltage and/or high current)

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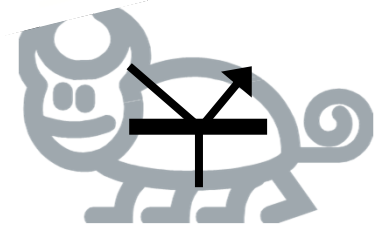
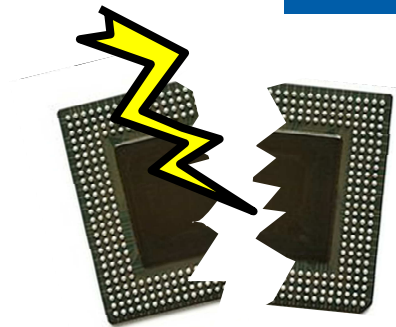
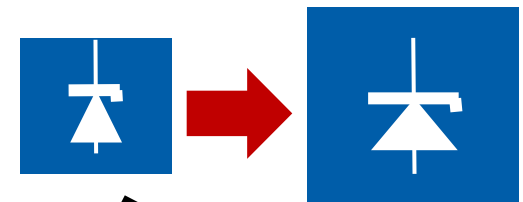
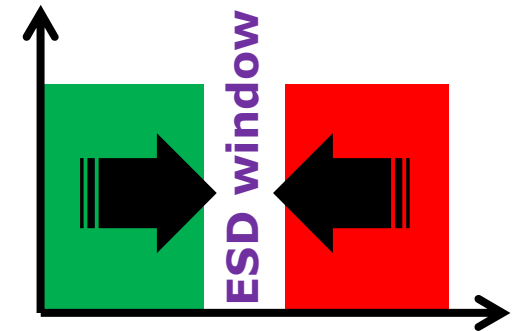
## HV technologies are typically used for:

- Power control circuits
- Power conversion circuits
- Power drivers (LED, LCD, PDP, Audio)
- Automotive applications
- Sensor and actuators driving circuits

# Challenges of HV design

## ■ Main challenges typical of HV technologies

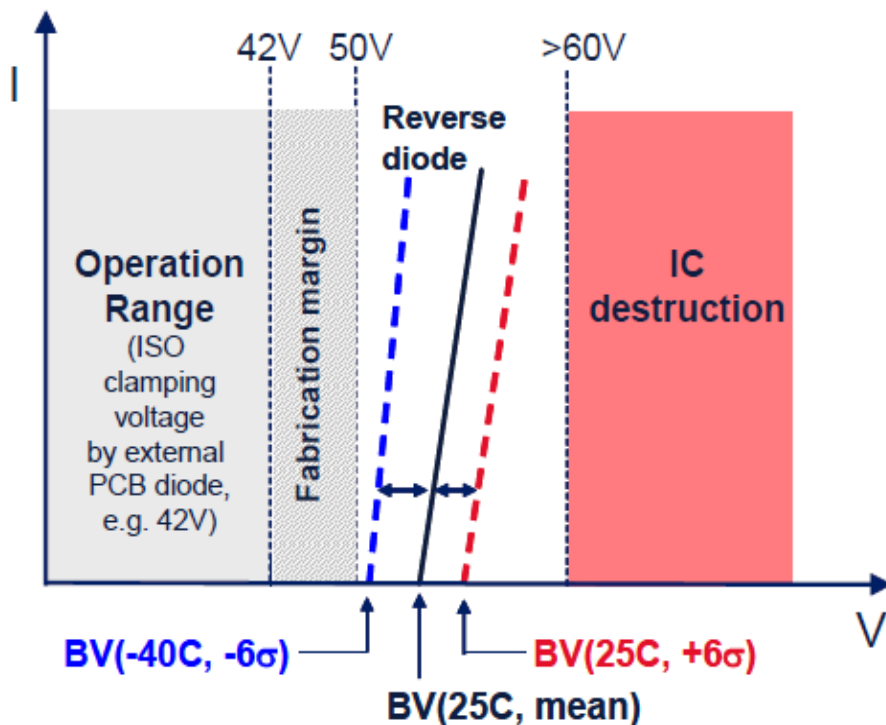
- Narrow ESD Design Window  
→ ESD window tends to vanish
- Large Area Solutions  
→ Higher clamping voltage requires larger device area
- Fragility of typical HV devices  
→ Ballasting is not effective like in CMOS technologies
- Increased risk of parasitic BJT (both NPN & PNP) possibly inducing LU issues



# ESD window concept in HV technologies

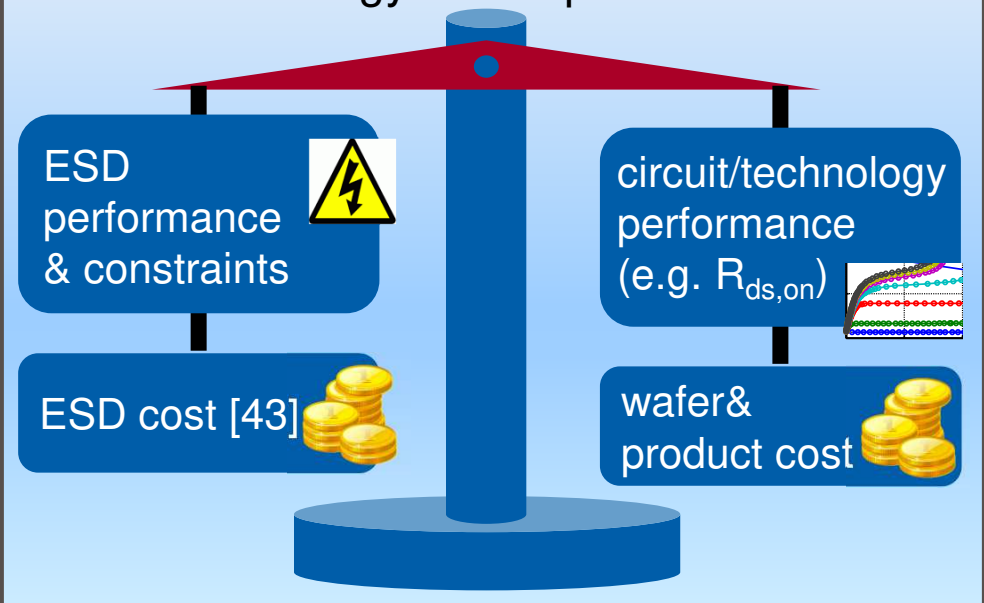
- Two concurrent trends in the HV world
  - Typical for HV MOS: trigger voltage ↓ , if  $V_{gs}$  ↑
  - External constraints ask for increasing voltage capability

→ ESD window tends to vanish

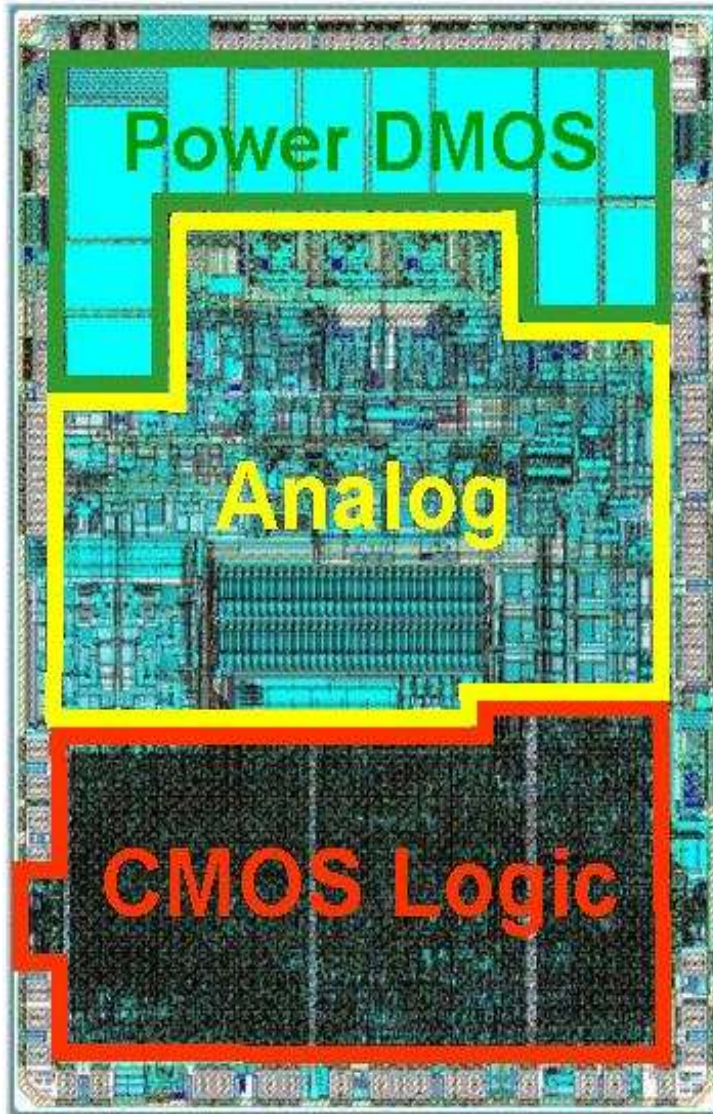


## Important considerations:

- ESD protection development concurrent with technology development

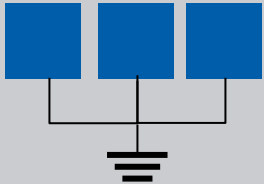
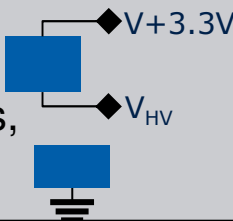
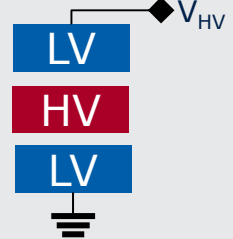


# Example of chip partitioning

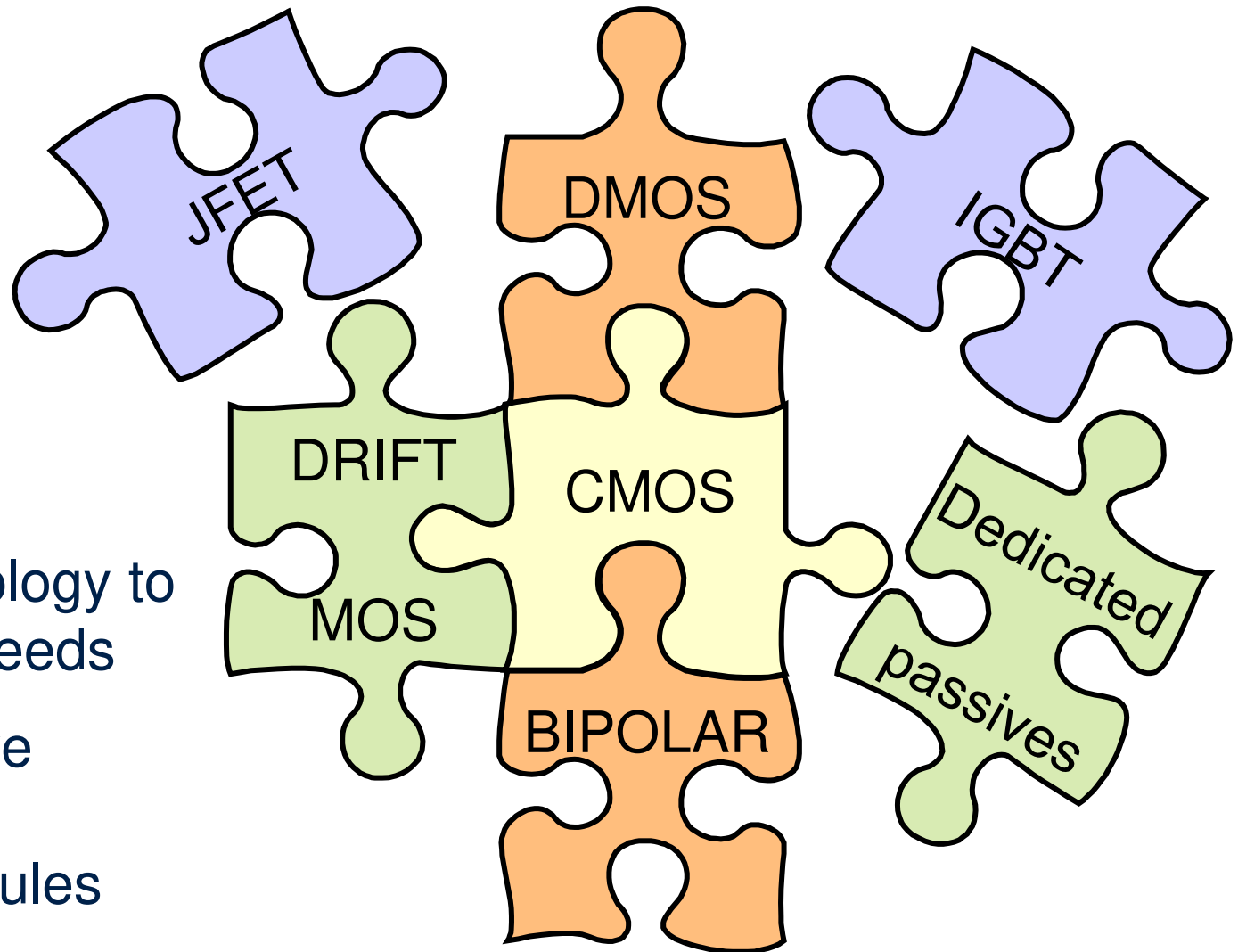


- DMOS:
  - Switches
  - Voltage regulator
- Analog:
  - Gate drivers
  - Diagnostics
  - Analog I/O
- CMOS:
  - Digital control
  - Digital I/O

# What is special for high voltage ICs?

| CMOS Integrated Circuits                                                                                                                                  | VERSUS | HV Integrated Circuits                                                                                                                                                    |
|-----------------------------------------------------------------------------------------------------------------------------------------------------------|--------|---------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| <p>Standard I/O cells &amp; libraries</p>                                |        | <p>Custom analog I/O cells</p>                                                         |
| <p>Supply voltage of different domains in limited range</p> <p> <math>V_{CC1} = 1.8V</math> <math>V_{CC2} = 1.8V</math> <math>V_{CCIO} = 3.3V</math> </p> |        | <p>Supply &amp; pin voltages of different domains vary largely</p> <p> <math>V_{CC} = 3.3V</math> <math>V_{sup} = 12-40V</math> <math>V_{clamp} = 60V</math> </p>         |
| <p>All circuit blocks ground referenced</p>                            |        | <p>Floating circuit blocks e.g. high side driver circuits, supply by charge pump</p>  |
|                                                                                                                                                           |        | <p>Analog circuits with LV devices in HV domain</p>                                  |

# HV technologies: a flexible world



## Modularity:

- Adapt technology to application needs
- Add & remove devices and process modules

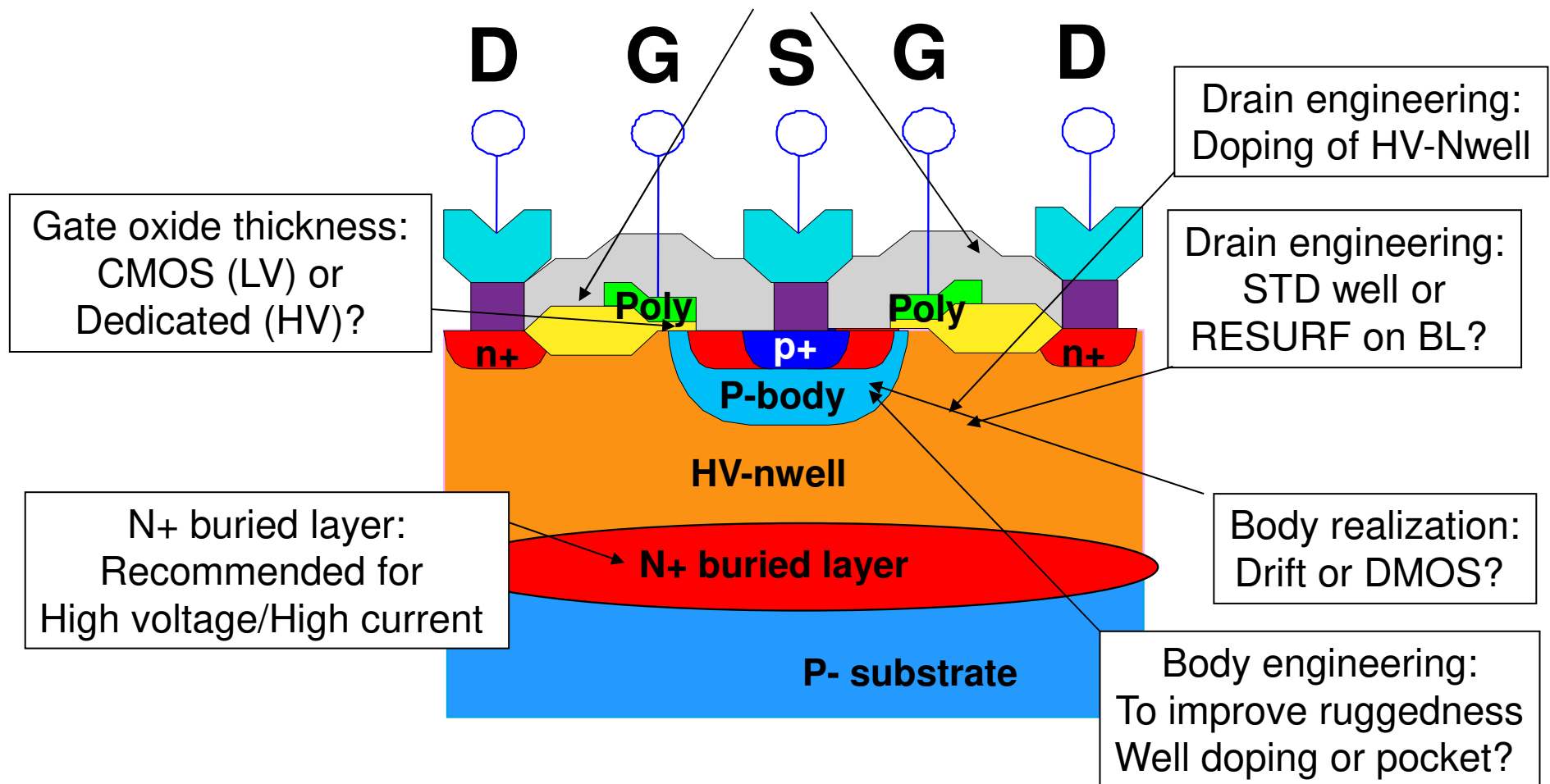
# BCD – Definitions & Motivations

- High Voltage technologies → commonly referred to as Smart Power or BCD technologies
  - B = Bipolar
    - Typically used for precise analog blocks
  - C = CMOS
    - Typically used for digital blocks
  - D = DMOS
    - Typically used for HV & Power stages
- New challenges in the Smart Power world
  - Analog design at low voltage & power levels → **Noise issues**
  - Power devices getting smaller → Increasing **power dissipation** by unit area

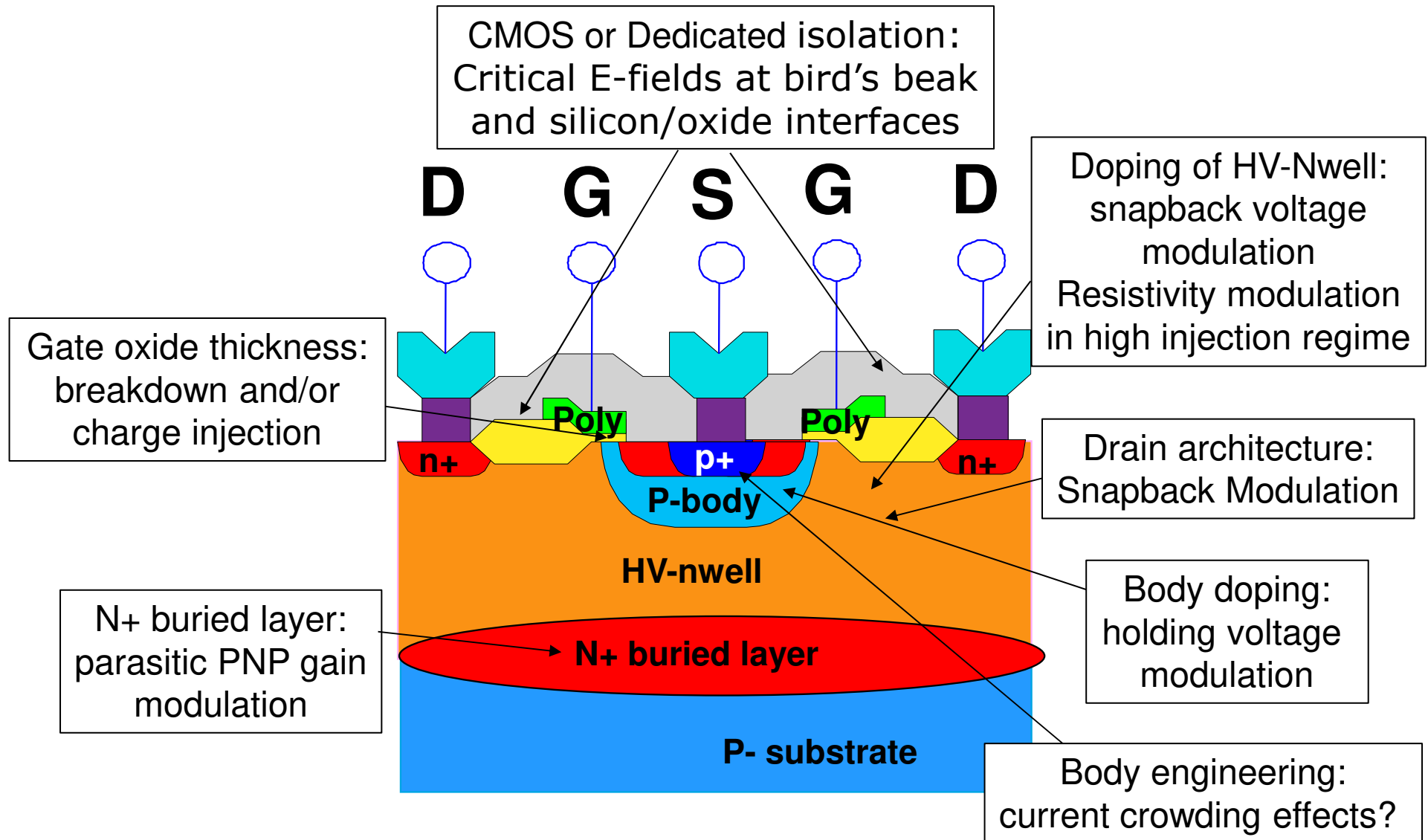


# Power-MOS: architecture engineering

CMOS or Dedicated isolation:  
(LOCOS vs. STI)  
For Field-Plate realization



# Power-MOS: the ESD performances



# Junction vs. Deep Trench Isolations

- HV transistors need to have high Breakdown voltage vs. substrate

## Junction Isolation

- Low doped region implemented to sustain the HV requirement

↑ Pro: No extra-steps required

↓ Con: Large area required

↓ Con: Parasitic lateral bipolar & MOS transistors to be considered carefully



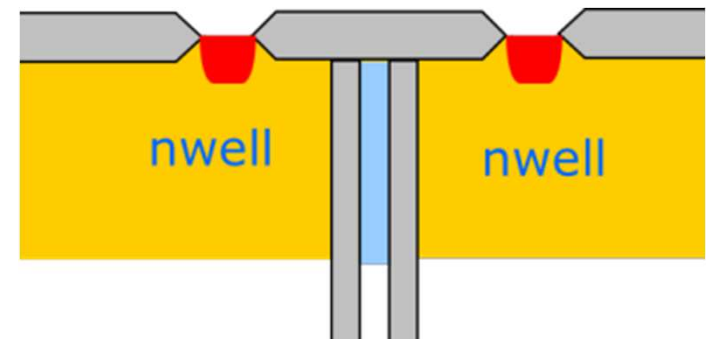
## Deep Trench Isolation

- High BV guaranteed by oxide-filled deep trench (depth ~ several um) surrounding the devices

↓ Con: Dedicated etch required

↑ Pro: More compact layout

↑ Pro: Better LU performances expected due to inhibition of parasitic lateral bipolars

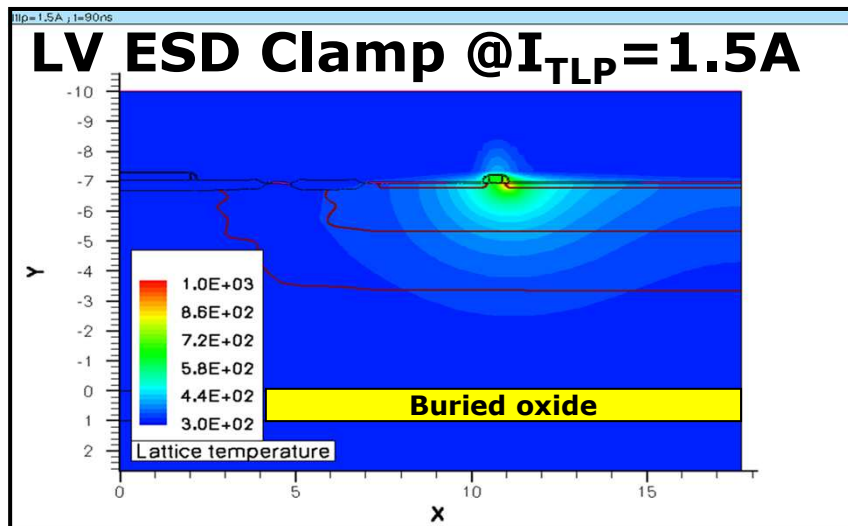
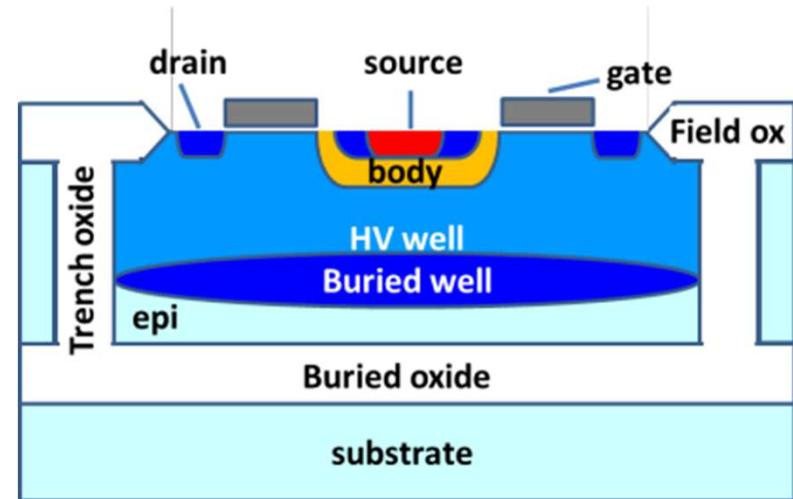


# HV SOI: Pros & Cons

- Different SOI approach adopted compared to CMOS → much thicker active Si layer epitaxy (2-10μm)

- Main pros

- ☐ bipolar parasitic inhibition
- ☐ HV devices (>100V) available



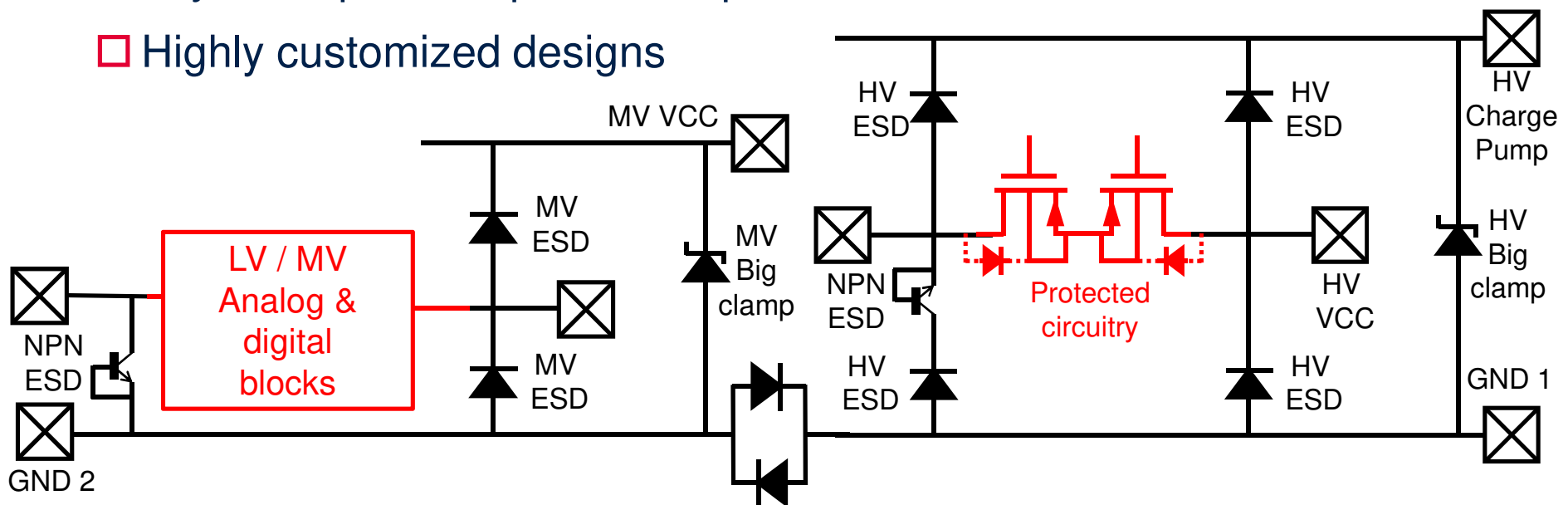
- ESD current flow and power dissipation confined in surface region → ESD performances not impacted
- EOS: increased power dissipation to be carefully considered

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# ESD networks in the HV world

- Typical complexity of an ESD network in the HV world: **very high**
  - Different power domains (different voltage ratings) **simultaneously** present
  - Analog circuits → possible uncontrolled alternative ESD paths
  - LV transistors used in HV domains
  - Layout-dependent parasitic bipolar transistors
  - Highly customized designs

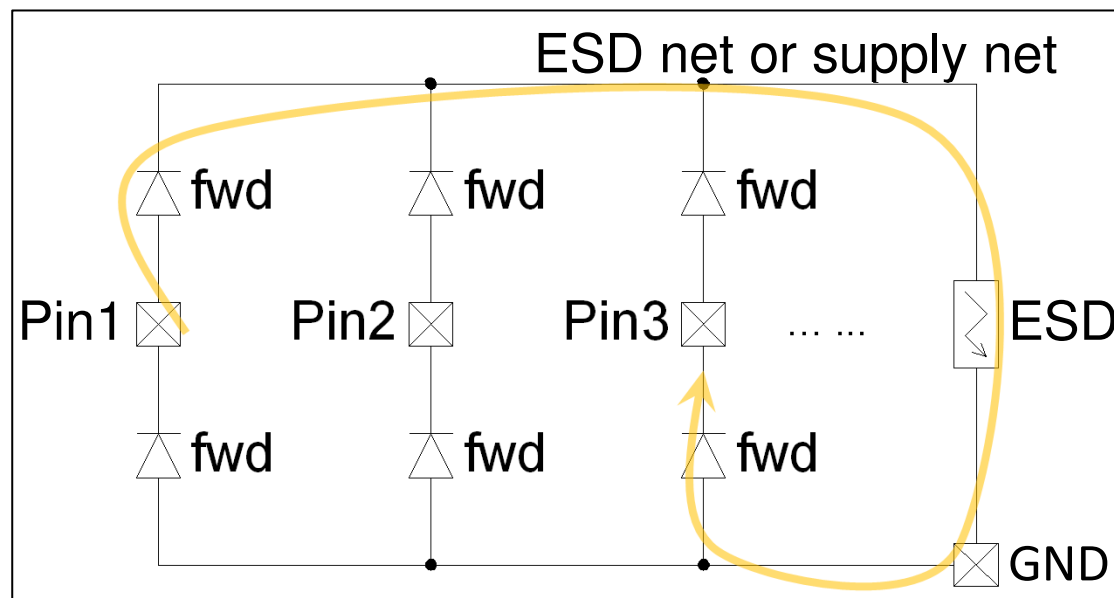


# ESD-Robust design flow in Smart Power ICs

- Standardized protection schemes can be adopted for logic I/O pins only
- Analog stages, negative pins, power blocks require a customized approach with ad-hoc solutions inside the protection network general concept
  - Deep knowledge of sensitive circuits is mandatory
  - Strict control of internal nodes must be pursued
  - Self-protected circuit blocks can be identified
  - Tolerance vs. parasitic transistors must be always double-checked
- Circuit simulation as a tool for ESD robustness prediction
  - Reliable for HBM on I/O circuits
  - Far less consolidated for CDM
  - Key to implement a robust ESD network for System-Level ESD requirements

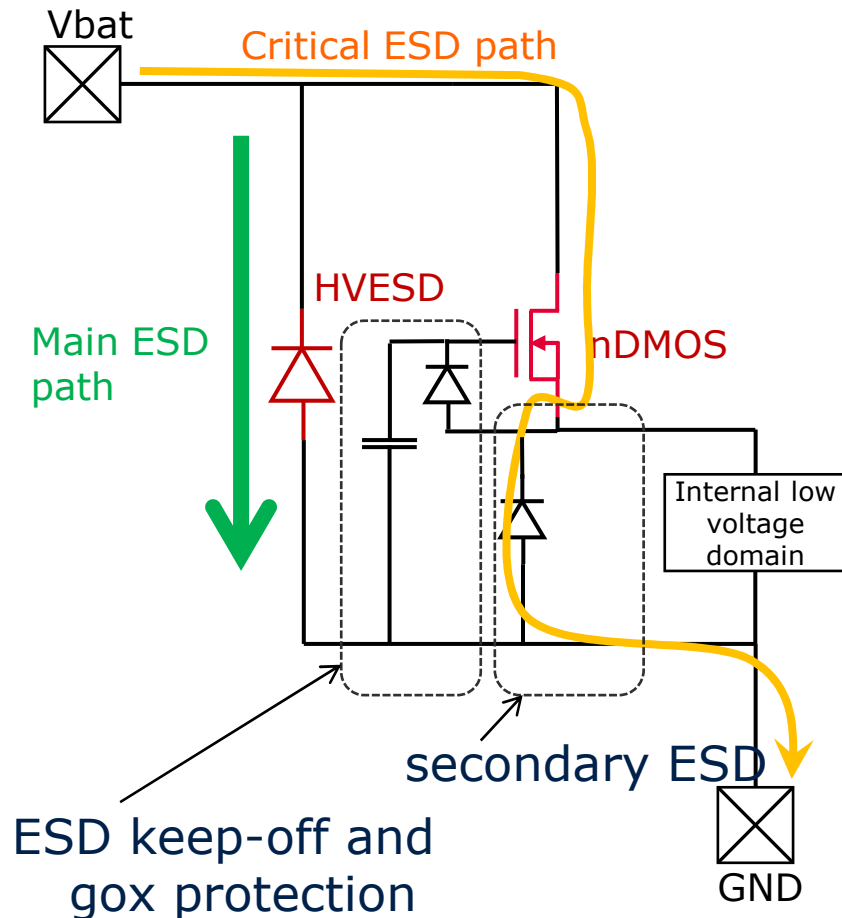
# General HV ESD Concepts /1

- Basic ESD concepts and networks which optimize area efficiency
- Be aware of forward recovery effect, low leakage floating node, metal interconnection routing, etc.



Dual-diode concept origin from CMOS can be also applied for HV ESD protection

# General HV ESD Concepts /2



- Voltage regulator involving mixed voltage ESD design: stack LV and HV devices
- nDMOS can pass significant ESD current to internal low voltage domain
- ESD keep-off e.g. capacitor suppress dynamic turn-on of nDMOS
- Secondary ESD protection required to protect internal domain

# HV protections approaches: Bipolar-based

- Several approaches already presented to built effective ESD protections based on breakdown or bipolar actions
- Typical solutions: NPN, SCR, PNP, diodes
  - ↑ Pro: Area effective
  - ↑ Pro: Local clamping possible → reduced impact of parasitic metal resistances
  - ← May need dedicated steps
  - ← Tight process control required (esp. during process transfer to different fabs)
  - ↓ Con: Reduced ESD window
  - ↓ Con: Deeply impacted by conductivity modulation → difficult to obtain triggering uniformity

# HV NPN & SCR /1

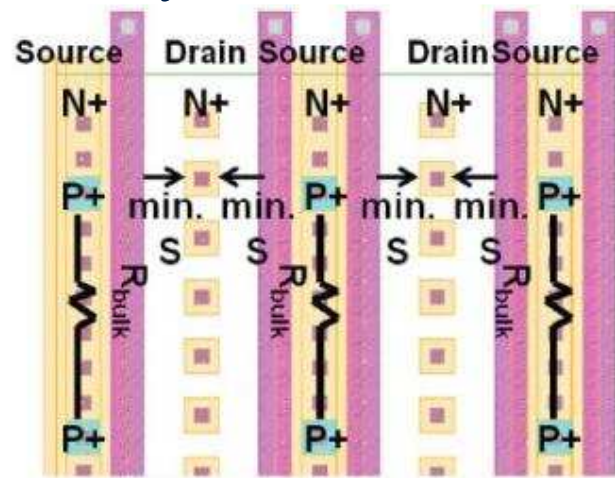
- Snapback-based (NPN & SCR) ESD protections show superior performances → higher area efficiency
- Main design issues of HV snapback protections
  - High conductivity modulation (Low doped regions)
  - High local E-field in singular points (e.g. bird's beak)
  - Isolated Body regions
    - Risk of early filamentation (non-uniform current flow)
  - Triggering voltage engineering **mandatory**
    - Risk of too high or unstable triggering (ineffective protection)
  - Holding voltage control
    - LU triggering risk if holding voltage < supply voltage (unacceptable in several application fields)

# HV NPN & SCR /2

- Several techniques proposed to improve current uniformity and snapback control

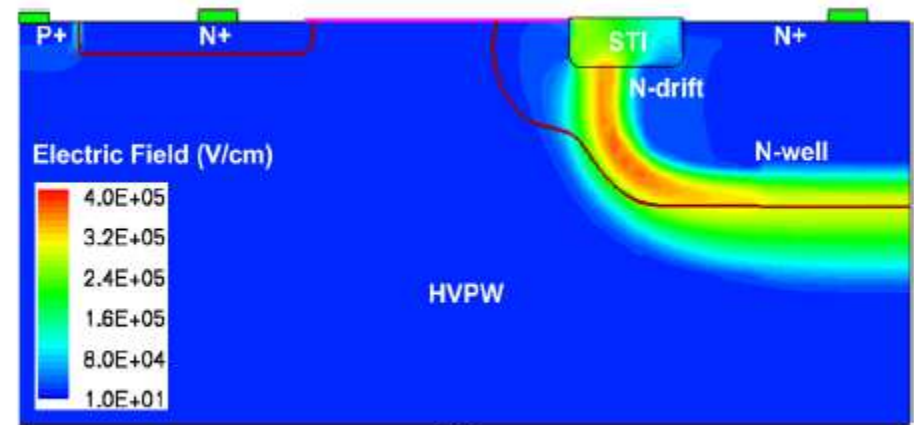
❑ Dedicated deep implants

□ Body resistance modulation

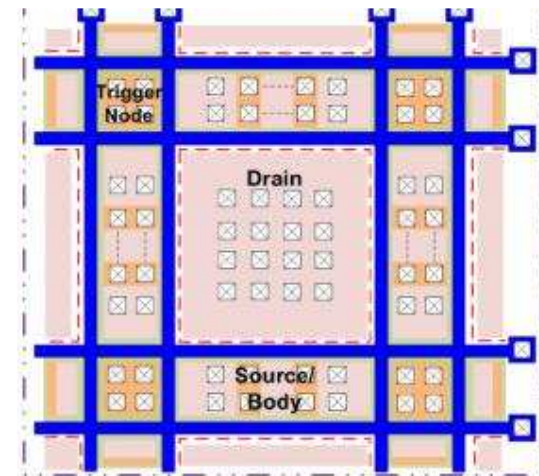


Courtesy of J. H. Lee [34]

- Triggering circuits embedding in HV transistor

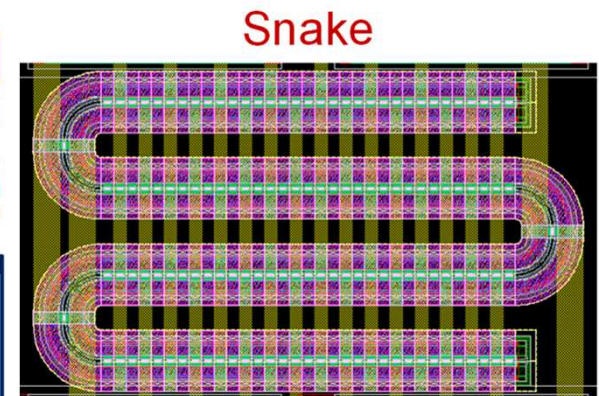
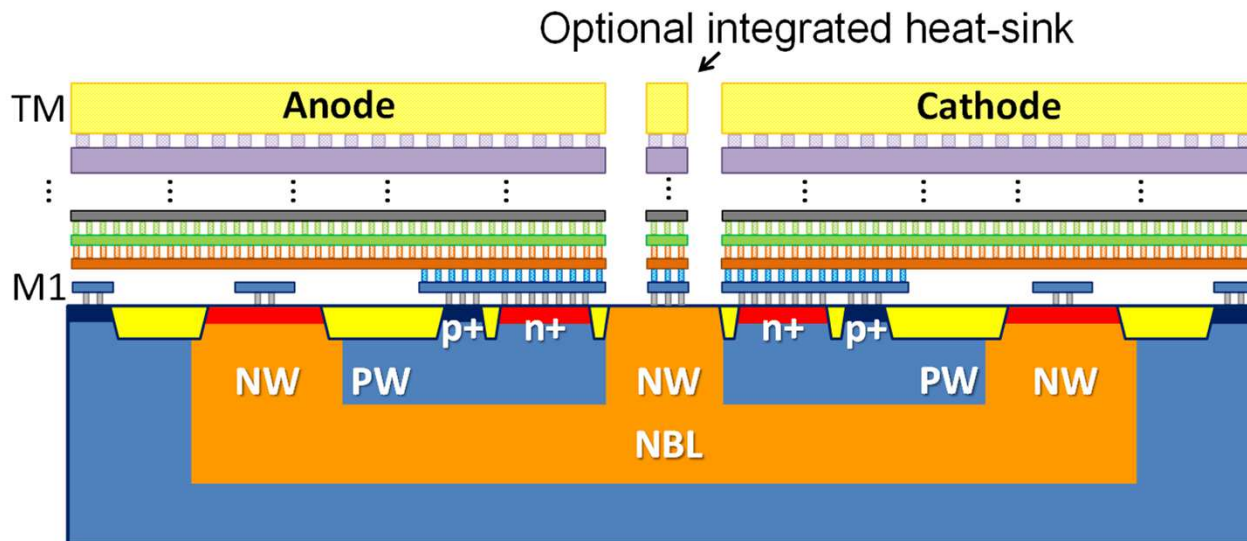


Courtesy of T. H. Lai [31] (b)



Courtesy of W. Y. Chen [12]

# BI-SCR - example

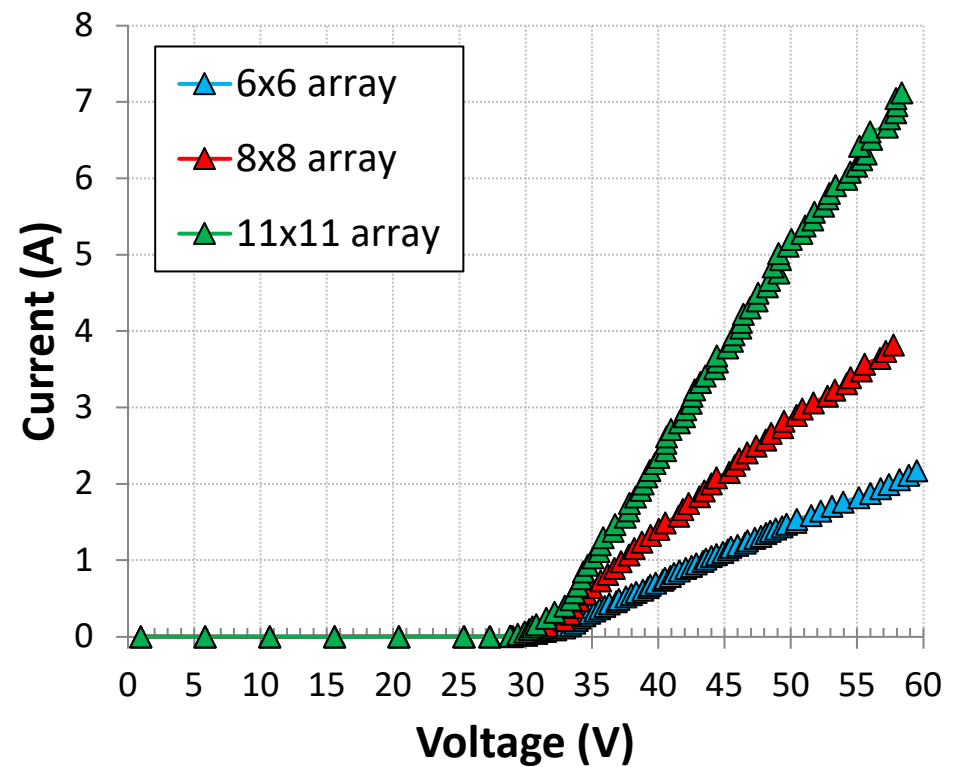


Courtesy of A. Tazzoli [73]

- Bi-directional SCR are often ad-hoc developed to implement on-chip system level protection cell
- Main characteristics and requirements:
  - Tunable Breakdown voltage
  - Low leakage (pA) & Low capacitance (pF)
  - Able to handle component- & system-level ESD stresses and also longer EOS (e.g. surge events)

# HV PNP

- Specific concerns defined by safety-driven issues (e.g. automotive):
  - holding voltage must be higher than the application operating voltage and max rating
- Suitable alternative to standard NPN → **PNP protections**
- Main advantages
  - No (or reduced) snapback
  - No ballast required
  - Matrix layout easily implementable
- Main limitations
  - Low  $R_{on}$  difficult to be obtained



# HV protections approaches: MOS-based

- High performing HV MOS already available in Physical Design Kit  
→ suitable to build effective protections
- Two possible triggering mechanisms: dynamically- vs. statically-activated MOS
  - ↑ Pro: Standard components used (no extra development effort)
  - Remote clamping implemented → parasitic metal resistances to be minimized
  - ↓ Con: Large area required

## ■ Dynamically-activated

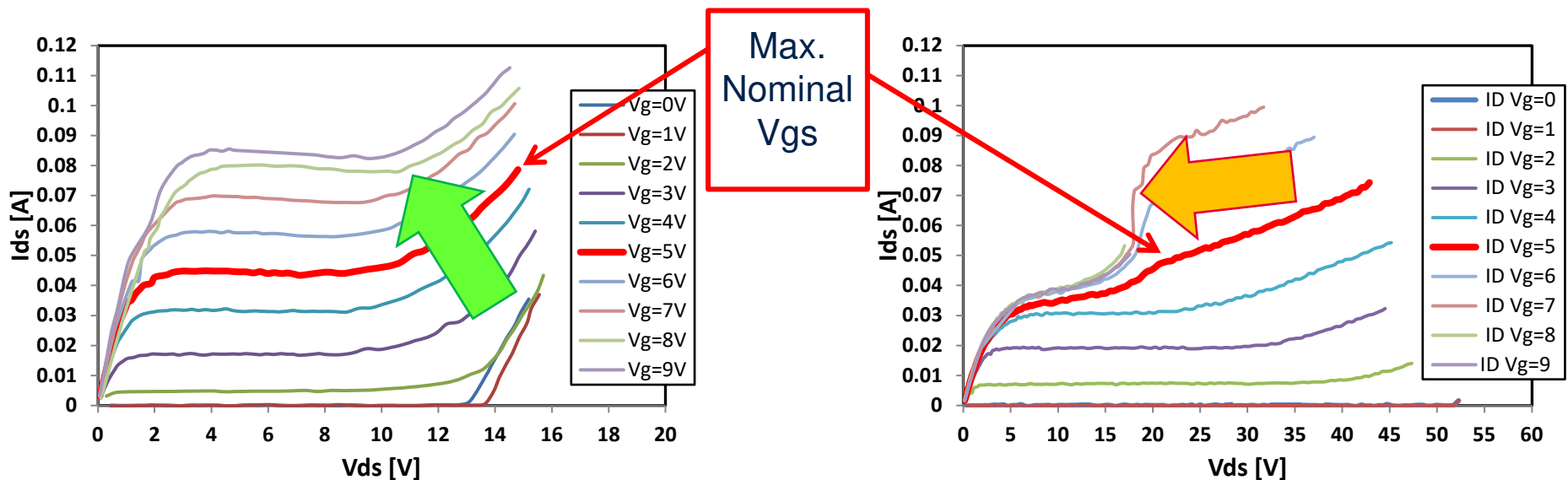
- ↑ Pro: less impacted by ESD window constraints
- ↓ Con: Possible interaction with device functionality and EMC testing

## ■ Statically-activated

- ↑ Pro: not prone to mistriggering due to  $dV/dt$  effects
- ↓ Con: Reduced ESD window for statically triggered clamps

# Active clamp Safe Operating Area

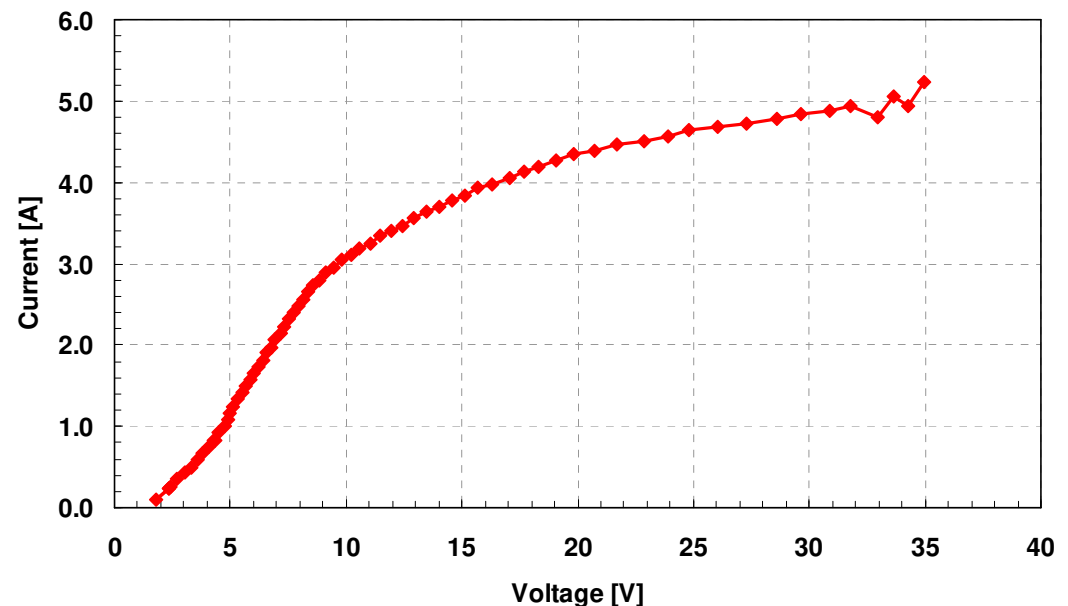
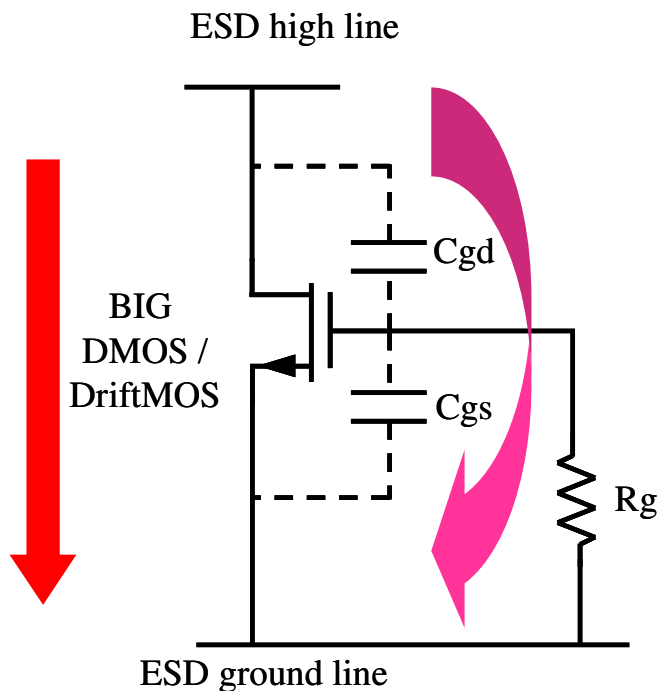
- HV MOS SOA data useful to optimize clamp performances
- Quasi-static IV curves at high  $V_{gs}$  &  $V_{ds}$  (not impacted by thermal effects) obtained with 100ns-TLP
- Dedicated ESD SOA can be different from standard operating conditions (defined by reliability concerns)



- ↑ Pro: overdrive beyond nominal condition can improve current capability
- ↓ Con: excessive gate voltage may induce premature snapback

# Dynamically activated HV clamps

- High Drain/Gate capacitance of HV DMOS/DRIFT MOS allows to build efficient dynamic clamps



- Easily tunable clamping voltage → low values achievable
- Gate coupling ratio dependent on HV-MOS implementation  
→ Typical  $V_{gs}/V_{ds}$  ratios (with pulsed  $V_{ds}$ ) in the range  $1/15 - 1/5$

# Dynamic clamps: design guidelines

## ■ Development targets of dynamic clamps

1. Limit drain ESD voltage (e.g. 70% of AMR)
  - To reduce IC voltage stress
  - To increase voltage margin for ESD network (diodes, parasitic metal resistance...)
2. Maximize ESD gate overdrive (> standard operating SOA)
  - High MOS current capability, minimized ESD area

## ■ Dynamic clamps reduce constraints on protected circuitry

- Strong optimization of active performances
- Special transistors (IOMOS) may not be needed

# Voltage triggering of active clamps

- Dynamically triggered active clamps main drawback: **unintended turn-on** in operative condition

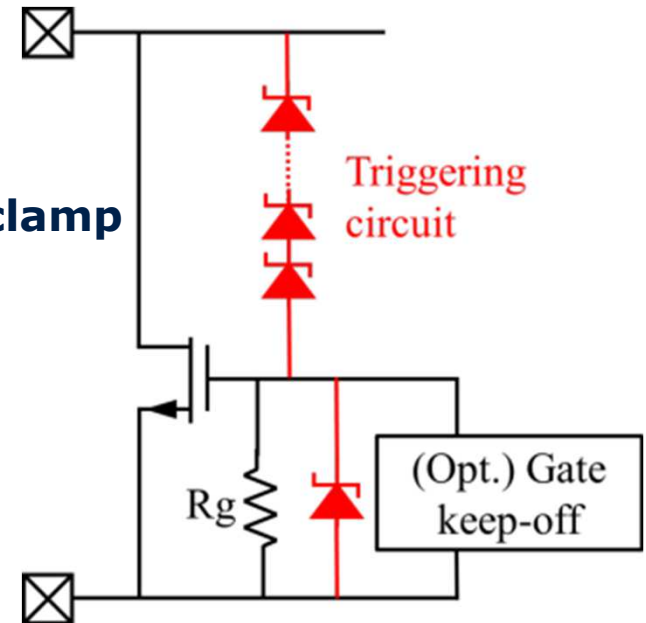
- Main root causes:

- ☐ Noisy supply lines caused by power stage switching
- ☐ EMC coupling of disturbs
- ☐ Hot plug of HV supply lines



- Statical trigger can be implemented with dedicated circuits (Zener chain...)
- Careful tailoring of the gate circuitry needed
  - ☐ Triggering circuit optimized to turn-on the clamp considering statistical process spread and voltage capability
  - ☐ Gate resistance → trade-off between disturb immunity (low  $R_g$  required) and suitably low current consumption (high  $R_g$  welcome)

**Type-A clamp**



# Simulations of HV active clamps

- Active clamps implemented with standard components
- No conduction in Breakdown and Snapback mode typically considered
- Two scenarios for circuit simulations:

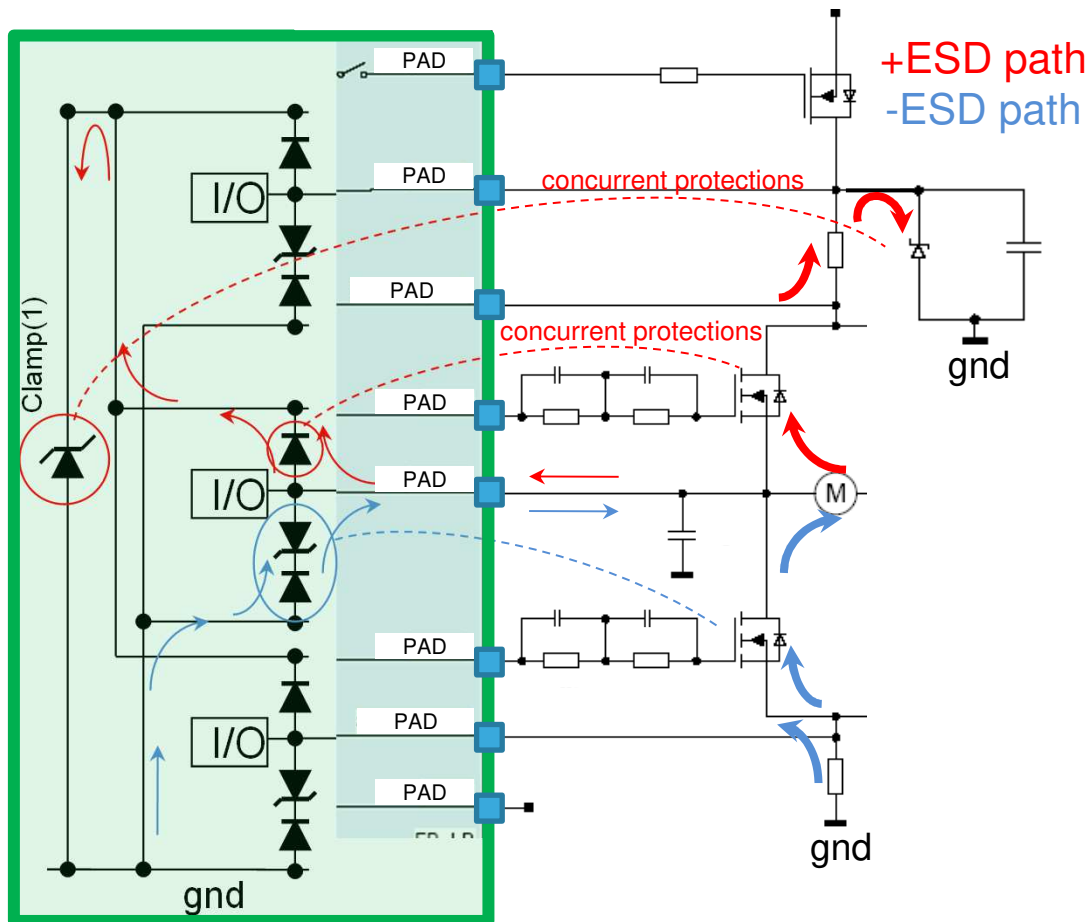
■ HV transistors are biased within the standard operating conditions

- ↑ Pro: No dedicated models needed
- ↑ Pro: No dedicated boundary conditions needed
- ↓ Con: Safe and conservative clamp implementation is possible

■ HV transistors are biased within the ESD SOA

- ↓ Con: Dedicated models needed including multiplication regions
- ↓ Con: Special boundary conditions needed to define safe regions
- ↑ Pro: Strong area optimization is possible

# Designing for System-Level ESD robustness /1



- External circuitry plays a major role
- ESD pulse may be clamped and/or distorted by external elements:
  - Discrete components (capacitors, PowerMOS...)
  - Board parasitics
- A simulation-based approach should be used to correctly design the best ESD solution for each pad

# Designing for System-Level ESD robustness /2

- The choice of the right ESD protection approach is key factor to ensure the right system-level robustness
  - Each on-chip protection should not by-pass the off-chip elements
  - Off-chip devices should work in their most effective working points
  - Each ESD approach has specific drawbacks to be considered

## Static-MOS based

- High clamping voltage may impact voltage capability of protected stages

## Dynamic-MOS based

- Low clamping voltage may impact off-chip protection (e.g. external capacitors show higher energy capability at high voltage)

## Snapback based

- Deep snapback may be risky for powered-on stresses

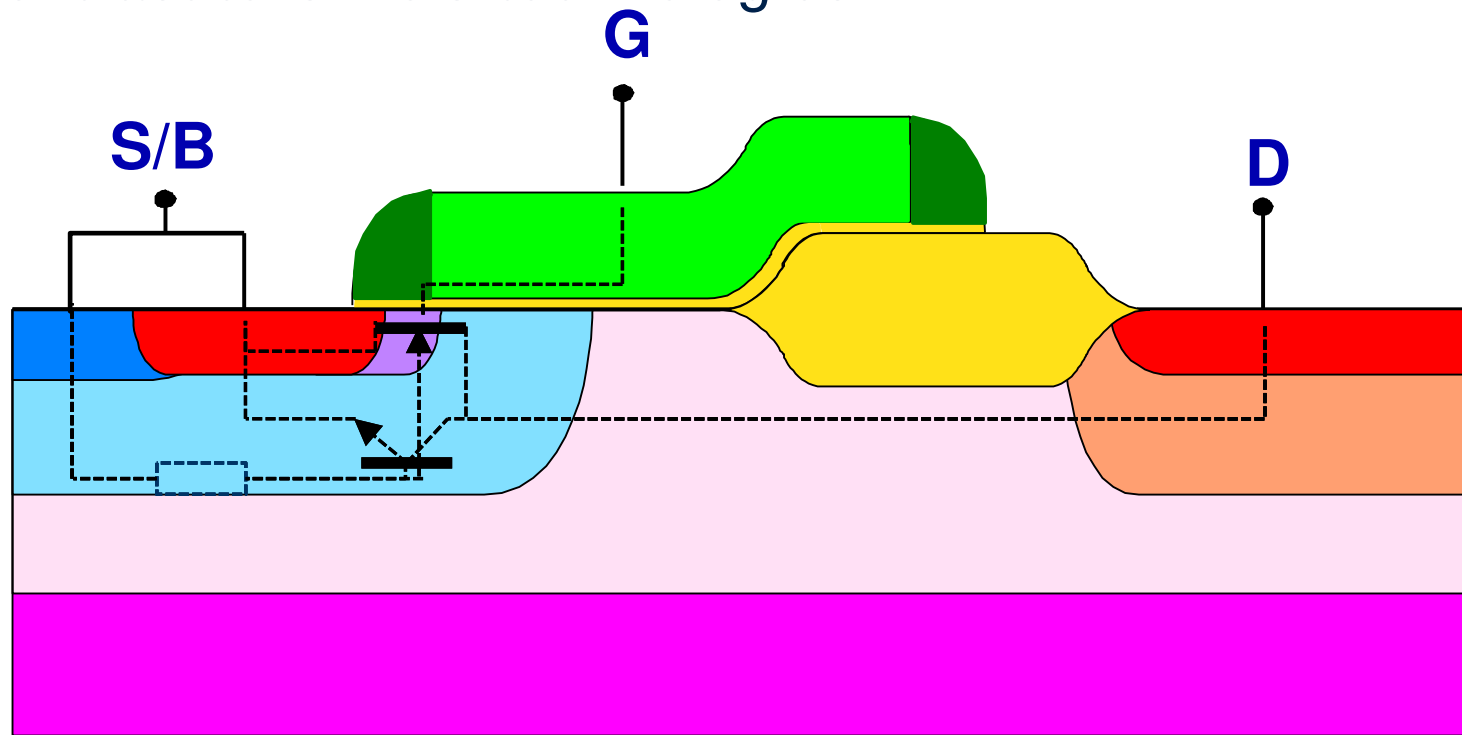
- There is no single and optimal solution fitting all requirements

# Contents

- Part 1 - Introduction: International ESD standards and norms
- Part 2 - Introduction to Smart Power Technologies & device portfolio
- Part 3 - ESD Protection concepts in Smart Power technologies
- Part 4 - HV related ESD issues

# Bipolar parasitics (device internal) /1

- N-LDMOS includes a parasitic bipolar NPN
- Triggering of this bipolar usually destroys the device
- Ballasting (*drain-contact-to-gate-space*) is not effective in contrast to CMOS technologies



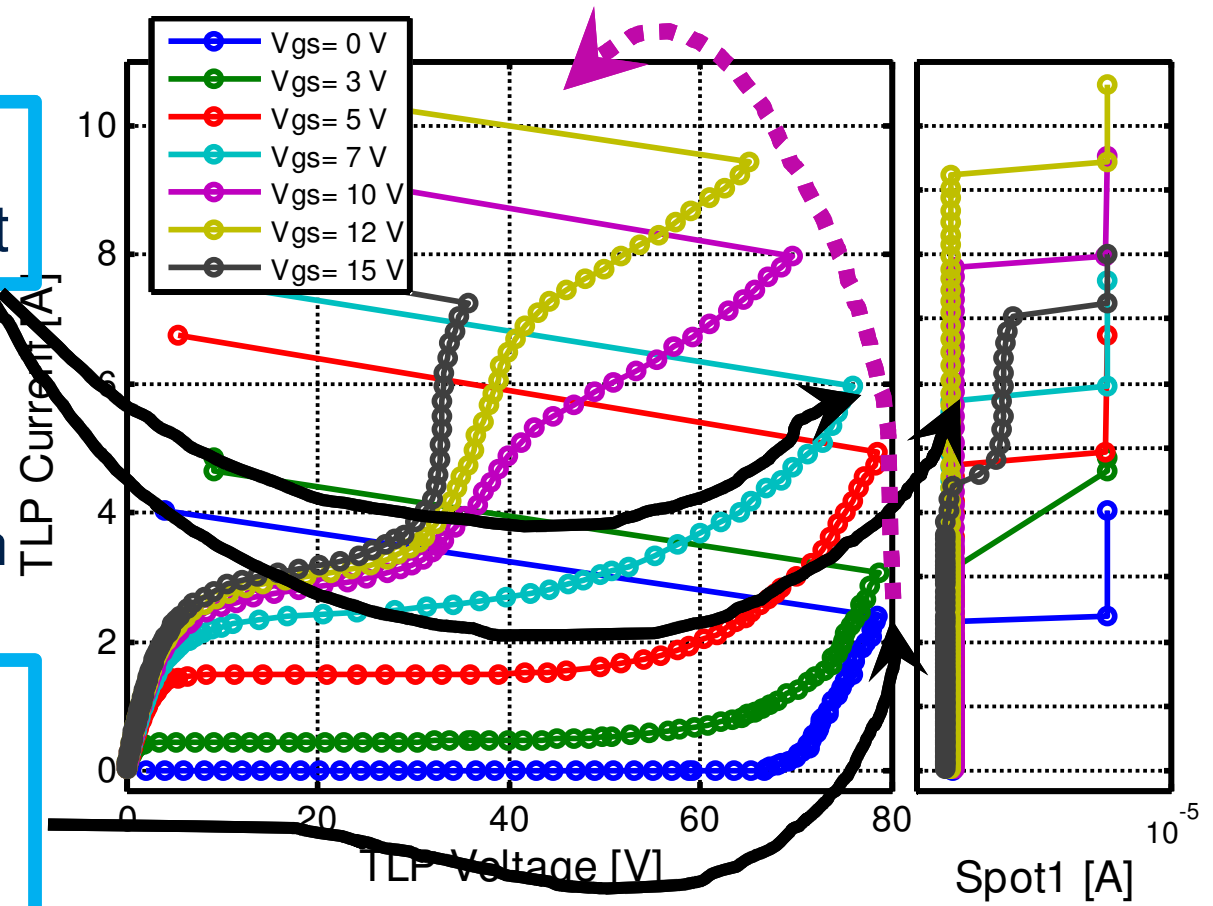
# Bipolar parasitics (device internal) /2

- Triggering of bipolar NPN leads to snapback

- Usually nLDMOS is not snapback-robust

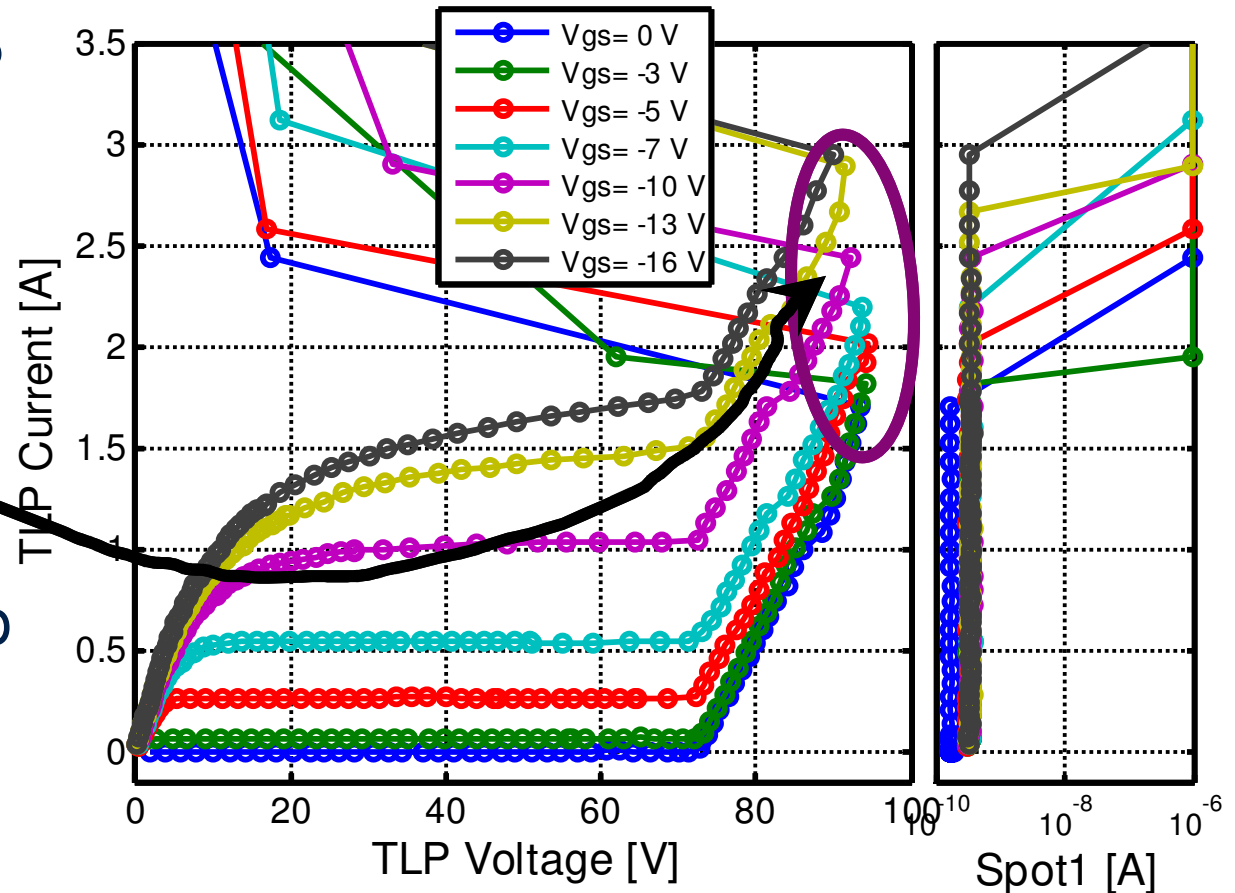
- Transient SOA characteristics: method to investigate device performance in detail

- Usually trigger point ( $V_{t2}$ ,  $I_{t2}$ ) depends on  $V_{gs}$



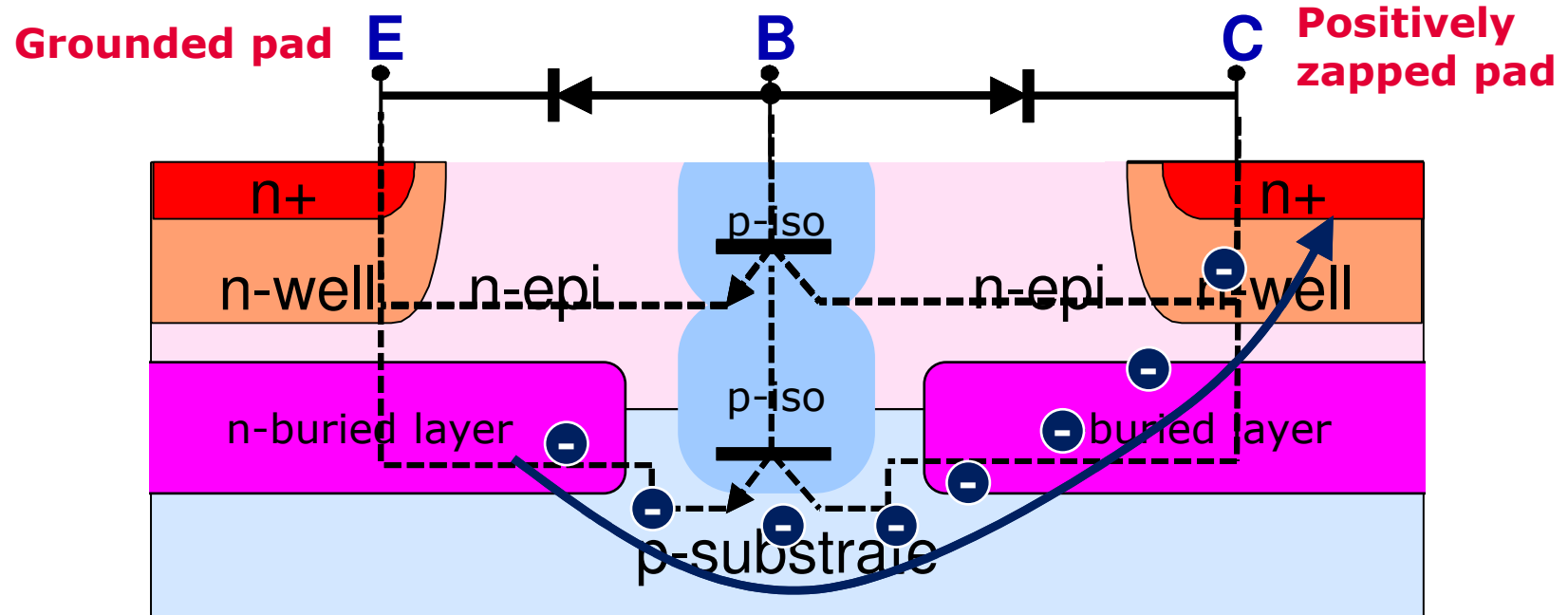
# Bipolar parasitics (device internal) /3

- PMOS devices have parasitic PNP
- Parasitic PNP is usually much more robust than NPN
- Less dependence on  $V_{gs}$
- PNP does not snap back as „easily“ as NPN



# Bipolar parasitics (between devices) /1

## ■ Cross-section of substrate NPN



- Epi-well below Ground → emitter
- Injection of minority carriers during ESD zap

- Epi well at high potential → collector
- Usually multiple collectors
- Snap-back risk of npn

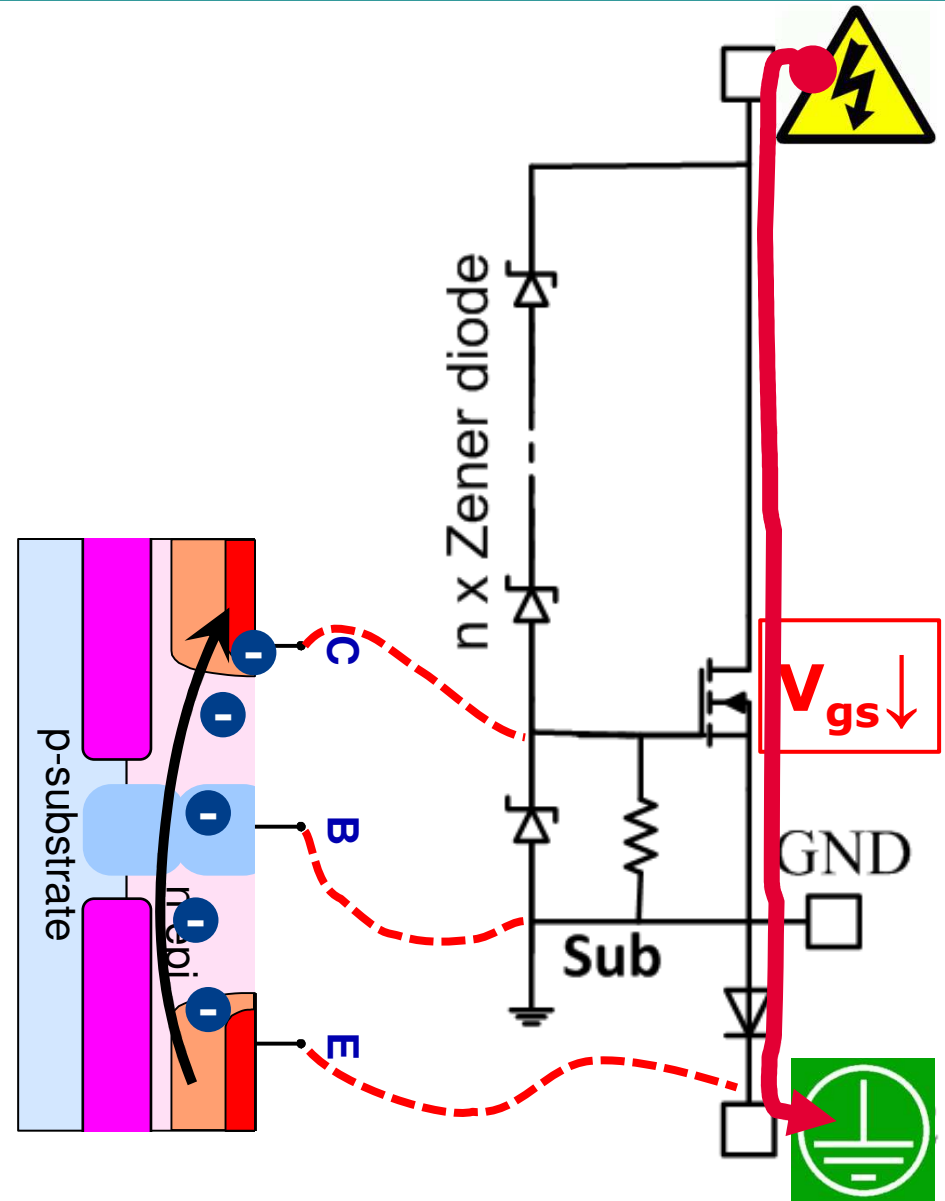
# Bipolar parasitics (between devices) /2

## ■ E.g. risk with active clamps:

- Collector node is connected to gate of clamp device
- Gate may be pulled down during ESD zap
- ESD clamping function fails

## ■ Majority carrier effects

- Substrate potential rise
- Active clamps may open due to forward bias of substrate / nwell diode (connected to gate) (unintended, in normal operation)





Thank you for your attention!

Now: Q & A time

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